

Serial No. 508,119
Filing Date 27 July 1995
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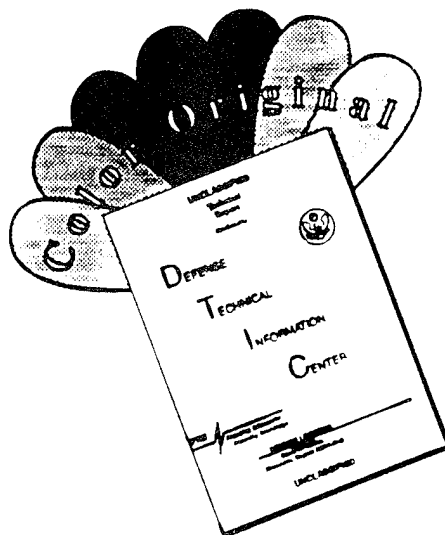
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Patent Application
NC 75,989

MULTI-OCTAVE, HIGH DYNAMIC RANGE OPERATION OF
LOW-BIASED MODULATORS BY BALANCED DETECTION

Field Of The Invention

5 The present invention relates to optical links and, more particularly, to an optical link operated by using a balanced detection technique so as to extend the bandwidth of the system, yet substantially eliminate even-order harmonic distortion of the modulated signal.

Background Of The Invention

10 In telecommunication systems, there are numerous applications in which it is preferred to transmit either analog or digital signals over an optical fiber instead of over an electrically conductive cable. Recently there has been a considerable development effort focused on improving the
15 microwave performance of fiber-optic links used in telecommunication systems. A major component involved in this developmental effort is an optical modulator.

20 Optical modulators, sometimes referred to as electro-optic modulators, receive coherent light in the form of a laser beam commonly referred to as an optical carrier. The optical modulator modulates the optical carrier with an informational signal having a relatively high frequency in the upper megahertz range. The optical modulator is commonly provided in the form of

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a Mach-Zehnder interferometer, well known in the art, and described, for example, in U.S. Patents 4,502,037; 5,109,441; 5,168,534; and 5,373,383; all herein incorporated by reference. The modulator serves as a waveguide and may be provided as an
5 integrated optic ("IO") device fabricated from a semiconductor or insulator material in a manner more fully described in U.S. Patent 5,168,534, as well as in the technical article "Linear Interferometric Waveguide Modulator for Electromagnetic-Field Detection" of C. H. Bulmer et al, published in OPTICS LETTERS,
10 Vol. 5, No. 5, May, 1980, and are herein incorporated by reference. The modulator provides an optical output which is commonly demodulated by a photovoltaic detector. The operation of the modulator and its associated detector circuit have certain drawbacks. More particularly, the modulator generates unwanted
15 harmonic components during its operation and the detector circuit has limitations with regard to its sensitivity.

FIRST PROBLEM

The Mach-Zehnder interferometers and the detector circuits are both desired to be operated in their linear range and which
20 is sometimes referred to as the linear dynamic range (LDR) of a communication link which is a measurement of the limits of operation of the link itself. The linear dynamic range of the Mach-Zehnder interferometer is of particular importance when

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serving as an electric field sensor. Improvements to this linear dynamic range are described in two (2) technical papers, with the first entitled "Increased Linear Dynamic Range by Low-Biasing the Mach-Zehnder Modulator" of M. L. Farwell et al, and the second
5 entitled "Maximum Dynamic Range Operation of a Microwave External Modulation Fiber-Optic Link" of E. Ackerman et al, both of which are also herein incorporated by reference.

As more fully described in the technical article of M. L. Farwell et al, Mach-Zehnder interferometers have a sinusoidal
10 response and are typically operated with a 50% optical bias, that is, the operating point is selected to correspond to the 90 degree ($\pi/2$) position on the response curve of the Mach-Zehnder interferometer, referred to as the quadrature point and which is also described in U.S. Patent 5,168,534 with reference to the
15 response curve or transfer function illustrated in Fig. 2 therein. The selection of the 50% optical bias, also commonly referred to as "Operating at Quadrature," is advantageous since it improves the linear dynamic range. Both of these technical articles describe further improvements to the linear dynamic
20 range by lowering the bias point to reduce the average DC level on the detector. Both of these low bias approaches allow for the use of laser sources having relatively high power available, while avoiding detector saturation. However, the operation of

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the modulator with decreased bias disadvantageously increases the unwanted second, and more generally even-order harmonic distortion of the modulator signal which limits the operation of the modulator to a narrow bandwidth. It is desired that the linear dynamic range of an optical link be increased, but without encountering increased unwanted even-order harmonic components so that the optical link may have a multi-octave response characteristic.

SECOND PROBLEM

Both the technical articles of M. L. Farwell et al and E. Ackerman et al describe that lowering of the bias point is accomplished without driving the detector circuit into saturation so that linear operation of the detector circuit is advantageously preserved. However, preserving the linear operation of the detector circuit while still allowing for the optical link to generate unwanted even-order harmonic components during its modulation operation is not a satisfactory solution. It is desired that the linearity of the detector circuit operation be preserved while at the same time eliminating the optical link's unwanted generation of the even-order harmonic components.

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Summary Of The Invention

Accordingly, an object of the present invention is to provide an optical link particularly suited for a telecommunication network that does not suffer from the prior art techniques of generating unwanted even-order harmonic components during its modulation operation.

Another object of the present invention is to provide for an optical link that cooperates with a linearly operated photovoltaic detector so as to improve the sensitivity therebetween and to more readily serve as an electric field sensor.

A still further object of the present invention is to provide an optical link having a predetermined bias so as to provide for a predetermined operating point that substantially eliminates the generation of unwanted even-order harmonic components, as well as provides for being able to respond to a relatively low-level applied stimulus so as to manifest a relatively high sensitivity particularly suited for electric field sensor applications.

Further still, it is an object of the present invention to provide an optical link that increases the linear dynamic range

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of a fiber optic telecommunication network.

The present invention is directed to an optical link that is operated using a balanced detection technique so as to extend the linear dynamic range of a telecommunication system, yet
5 substantially eliminate even-order harmonic distortion of the modulated signal.

The optical link modulates an optical carrier with an information signal and comprises an optical power splitter, first and second modulators, first and second detectors and a 180°
10 phase shift device. The optical power splitter receives the optical carrier and provides about one-half of the power of the optical carrier to a first path and the remaining power of the optical carrier to a second path. The first modulator receives the optical carrier having about one-half of the original power
15 at the first path and has means for receiving both the information signal and a steady-state bias signal with a positive offset quantity to preset the operating point of the first modulator. The first modulator provides a first optical output signal modulated by the information signal and representative of
20 the optical carrier present at the first path. The second modulator receives the optical carrier having about one-half of the original power at the second path and has means for receiving

both the information signal and a steady-state bias signal with a negative offset quantity to preset the operating point of the second modulator. The second modulator provides a second optical output signal modulated by the information signal and
5 representative of the optical carrier present at the second path. The first detector receives the first optical output signal and develops a representative first rf output signal therefrom. The second detector receives the second optical output signal and develops a representative second rf output signal therefrom. The
10 180° phase shift device is connected to receive one of the first and second rf output signals and has means for connecting its output to the other of the first and second rf output signals so as to provide for a combined output comprising the first and second output signals. The combined output serves as the output
15 signal of the optical link.

Brief Description Of The Drawings

These and other objects, features and advantages of the present invention, as well as the invention itself, will become better understood by reference to the following detailed
20 description when considered in conjunction with the accompanying drawings, wherein like reference numbers designate identical or corresponding parts throughout the several views, and wherein:

Fig. 1 is a block diagram of a prior art low-biased optical link.

Fig. 2 is a graph illustrating the operating point of the optical link of Fig. 1.

5 Fig. 3 is a block diagram of the optical link of the present invention having two parallel arranged Mach-Zehnder interferometric modulators.

Fig. 4 illustrates the operating points for each of the Mach-Zehnder interferometers arranged in Fig. 3.

10 Fig. 5 is composed of Figs. 5(a) and 5(b) that respectively illustrate the prior art waveforms of the fundamental component and the unwanted second-order harmonic component of an information signal used in modulating an optical carrier.

15 Fig. 6 is composed of Figs. 6(a) and 6(b) that respectively illustrates the waveforms of the present invention of the fundamental component and the reduced unwanted second-harmonic component of an information signal used in modulating an optical carrier.

Fig. 7 illustrates performance parameters associated with the prior art optical link of Fig. 1.

Fig. 8 illustrates the improved performance parameters associated with the optical link of the present invention.

5 Detailed Description Of The Preferred Embodiments

With reference to the drawing, Fig. 1 illustrates a prior art optical link 10 which may be first described in order to more fully appreciate the principles of the present invention. The optical link 10 embodies the operational parameters described in both the technical articles of E. Ackerman et al and M. L. Farwell et al.

The optical link 10 receives coherent light I_{IN} from a laser 12, and comprises an amplitude modulator 14, such as a Mach-Zehnder interferometer, and a detector 16. The laser 12 is preferably connected to the modulator 14 by an optical fiber 18, well known in the art, which carries the optical power signal I_{IN} generated by the laser 12. The output I_{OUT} of the Mach-Zehnder interferometer 14 is coupled to the detector 12 by an optical fiber 20.

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The laser 12 may be an Nd-YAG laser having an operating wavelength of 1.3 micrometers and having a power output of about 300 milliWatts (mW). The present invention, as to be described, makes use of this relatively high available power of the laser 12
5 for driving the light sensing circuit of the optical link of the present invention.

The Mach-Zehnder interferometer 14, generally referred to as an amplitude modulator, is well known and may be of the type disclosed in the previously incorporated by reference U.S.
10 Patents 4,502,037; 5,109,441; 5,168,534; and 5,373,383 and technical article of C. H. Bulmer et al. The operating principles of the Mach-Zehnder interferometer are well known and need not be further described herein but, if desired, reference may be made to these incorporated by reference U.S. Patents and
15 the technical article of C. H. Bulmer et al for further description thereof.

The detector 16 may be an InGaAs p-i-n photodiode operated in its linear region and more particularly, operated to avoid its saturation or non-linear operational condition, as more fully
20 described in the technical article of E. Ackerman et al. The detector 16 provides an electrical output signal P_{Ac} which is the output of the optical link 10 of Fig. 1.

The operation of the modulator 14 is primarily controlled by a signal 22 applied to an electrode 14A of the modulator 14. The signal 22 comprises an rf signal, serving as the information signal, and a phase bias signal ($\pi - \alpha$) which is given in radians and is obtained by applying an appropriate DC voltage. The information signal may have a typical fundamental frequency of 110 MegaHertz (Mhz). The phase bias signal ($\pi - \alpha$) determines the operating point of the modulator 14 and may be described with reference to Fig. 2.

Fig. 2 has an X axis indicating the bias signal, given in radians achieved by applying a DC voltage to electrode 14A, and a Y axis indicating the amplitude of the modulator optical output signal I_{OUT} given in milliwatts. The X axis is further indicated to show the phase relationship of the signal I_{OUT} , wherein the 180° phase position is indicated as π . Fig. 2 further illustrates that a steady-state DC bias voltage corresponding to a phase bias of $\pi/2$ establishes an operating point 24, and a steady-state DC bias voltage corresponding to phase bias of $\pi - \alpha$, where $\alpha < \pi/2$ establishes an operating point 26. The operating point 24 represents the 50% optical bias for the modulator 14 and establishes the "Operating at Quadrature" previously discussed in the "Background" section. The operating point 26 determines the operational response of the Mach-Zehnder interferometer 14 of

Fig. 1. Furthermore, the operating point 26 corresponds to the decreased bias point also discussed in the "Background" section which, as more fully disclosed in each of the technical articles of M. L. Farwell et al and E. Ackerman et al, causes the Mach-Zehnder interferometer 14 to provide for an increased linear dynamic range (LDR) but, as also previously discussed, disadvantageously generates unwanted even-order harmonic components. The present invention not only increases the linear dynamic range (LDR) of its modulator but also eliminates the generation of unwanted even-order harmonic components, and may be further described with reference to Fig. 3.

Fig. 3 illustrates an optical link 28 of the present invention comprising an optical power splitter 30 that receives the coherent optical carrier I_0 from laser 12, and which provides optical output signals I_{01} , and I_{02} , respectively to first and second modulators 32 and 34, each of which preferably comprises a Mach-Zehnder interferometer. The first modulator 32 has an electrode 32A which receives a signal 36 similar to the signal 22 described with reference to Fig. 1, with the exception that the steady-state DC bias voltage therein establishes operating point of the first modulator 32 at phase bias of $(\pi + \alpha)$, to be further described herein. Similarly, the second modulator 34 has an electrode 34A which receives a signal 38 which is the same as the

signal 22 already described with reference to Figs. 1 and 2 and establishes the operating point of the second modulator 34 at phase bias of $(\pi - \alpha)$.

The first modulator 32 provides an output optical signal I_1 ,
5 which is routed to a detector 40 by means of an optical fiber 42. Similarly, the second modulator 34 provides a second optical output signal I_2 that is routed to detector 44 by an optical fiber 46 and, preferably, by a piezoelectric fiber stretcher 48 to be further described. Both detectors 40 and 44 may be similar
10 to the detector 16 of Fig. 1, whereas the optical fibers 42 and 46 may be similar to the optical fiber 20 of Fig. 1.

The output of detector 40, an electrical output signal, is applied to a 180° phase shifter 50 which, in turn, has its output applied to a signal combiner 52, and the output of detector 44,
15 also an electrical output signal, is applied directly to the signal combiner 52. Alternatively, the 180° phase shifter 50 may be swapped and placed in the output path of detector 44, or phase shifters (not shown) may be placed in the output paths of both detectors 40 and 44 so long as the signals provided to combiner
20 52 are 180° out of phase. The piezoelectric fiber stretcher 48 may similarly be swapped and placed in the input path of detector 40 or fiber stretcher may be placed in the input paths of both

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detectors 40 and 44. Alternatively, the phase shifter 50 and combiner 52 may be combined into a single device (not shown) form the signal P_{AC} located at the output stage of the optical link 28.

5 The optical power splitter 30, located at the input stage of the optical link 28, is known in the art and receives, via the optical fiber 18, the optical carrier I_0 and provides about one-half of the power of the optical carrier I_0 to a first path having a predetermined path and the remaining power of the
10 optical carrier I_0 is directed to a second path, also having a predetermined path. The optical signal delivered to the first path is I_{01} and the optical signal delivered to the second path is I_{02} .

 The first modulator 32 receives the optical signal I_{01} and
15 modulates it with the information signal contained in input signal 36 and delivered to electrode 32A. The modulation of the optical signal I_{01} by the information signal is in accordance with the operating point established by the steady-state phase bias of $(\pi + \alpha)$. The second modulator 34 receives the optical
20 signal I_{02} and modulates that signal with the information signal contained in input signal 38 and delivered to electrode 34A. The modulation of the optical signal I_{02} by the information signal is

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in accordance with the operating point established by the steady-state phase bias of $(\pi - \alpha)$.

The first and second detectors 40 and 44 respectively receive the first optical output signal I_1 from the first modulator 32 and the second optical output signal I_2 from the second modulator 34 and provide representative electrical rf output signals therefor. The detectors 40 and 44 each preferably comprises a filtering network that reduces or even eliminates the DC signal components that may be present in optical output signals I_1 and I_2 .

The 180° phase shift device 50 is connected to receive one of the first and second rf output signals from detectors 40 and 44 and cooperates with the signal combiner 52 so that the first and second output signals from detectors 40 and 44, respectively, are electrically combined to form the output signal P_{AC} of the optical link 28. The operating points that control the modulation operation of the first and second modulators 32 and 34 may be further described with reference to Fig. 4.

Fig. 4 is similar to Fig. 2 and illustrates the operating point 26 already described for the modulator 14 of Fig. 1 and which is also the operating point 26 of the second modulator 34.

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However, in addition, Fig. 4 illustrates a further operating point 54 established by the phase bias ($\pi + \alpha$). Accordingly, the optical link 28 of Fig. 3 is provided with two operating points, the first being operating point 26 for the second modulator 34 and the second being operating point 54 for the first modulator 32.

In general, the operation of the optical link 28 uses a balanced detection technique to eliminate unwanted even-order harmonic distortion and also to provide for an improvement in its sensitivity. The term "balanced detection" technique refers to the combination of two signals so as to eliminate any unwanted components. The optical link 28 modulates the coherent carrier signal I_0 , in particular, the signals I_{01} and I_{02} present on the first and second paths, respectively, with the information signal, shown as an rf signal, that is contained in both input signals 36 and 38 and which may be represented by the following expression:

$$\phi_m \cdot \sin(\omega t) \quad (\text{Expression 1})$$

The quantity α that is used to establish the operating points 54 and 26 for the first and second modulators 32 and 34, respectively, preferably has a value of about $\pi/4$ or less. The

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first and second modulators 32 and 34 respectively receive the signals I_{01} and I_{02} and correspondingly provide output signals I_1 and I_2 with optical intensity that may be respectively expressed by the below expressions:

$$I_1 = \frac{1}{2} I_{01} [1 - \cos \alpha + \phi_m \sin \alpha \sin \omega t - (\frac{\phi_m}{2})^2 \cos \alpha \cos 2\omega t + \dots] \quad (2)$$

$$I_2 = \frac{1}{2} I_{02} [1 - \cos \alpha - \phi_m \sin \alpha \sin \omega t - (\frac{\phi_m}{2})^2 \cos \alpha \cos 2\omega t + \dots], \quad (3)$$

5 where $I_0 = I_{01} + I_{02}$ is the intensity of the carrier light I_0 .

A Bessell function expansion was used in the derivation of expressions (2) and (3). Because the quantity ϕ_m is typically relatively small, the higher order harmonic terms of expressions (2) and (3) are omitted. A comparison between expressions (2)
10 and (3) reveals that the terms are similar except for the quantities $+\phi_m \sin \alpha \cdot \sin(\omega t)$ and $-\phi_m \sin \alpha \cdot \sin(\omega t)$.

The signal I_1 of expression (2) is routed to the detector 40, and the signal I_2 of expression (3) is routed to detector 44. At each of the detectors 40 and 44, the DC component of the modulator signal I_1 and I_2 is blocked and the rf output signal of detector 40 is routed to the 180° phase shift device 50 which, in turn, provides an rf output that is applied to the signal combiner 52. The signal combiner 52, in one embodiment, combines the rf output signal from the 180° phase shifter 50 with the rf output signal of the detector 44 to form signal P_{AC} . In another embodiment, the 180° phase shifter 50 is in the output path of detector 44 but the combined signal P_{AC} remains the same. The combined electrical output signal P_{AC} of the optical link 28 has intensity represented by the following expression:

$$P_{AC} = \frac{1}{4} \cdot R \cdot I_0 [2\phi_m \sin \alpha \sin \omega t + \text{higher order odd harmonic components}], \quad (4)$$

where R is a factor representing the responsivity of detectors 40 and 44.

A review of expression (4) reveals that output signal P_{AC} is expressed in terms of I_0 , which is accomplished by assuming $I_{01} = I_{02}$ each of which is equal to $I_0/2$. A further review of expression (4) reveals that it is devoid of any even-order harmonic components. It should be noted that the elimination of the second order harmonic components (sometimes described herein)

also eliminates other even-order harmonic components.

Accordingly, the operation of the optical link 28 modulating the optical carrier, I_0 , with the information signal $\phi_m \sin(\omega t)$ provides for the substantial elimination of the even-order harmonic components.

In order to effectively provide for the elimination of the even-order harmonic components created by the combining of the signals I_1 and I_2 and phase shifting one of the signals I_1 and I_2 with respect to the other, the power contained in the even-order harmonics of the separate signals I_1 and I_2 output from modulators 32 and 34 should be substantially equal. One method for providing these signals I_1 and I_2 with equal power is to provide an adjustable optical power splitter 30 so as to compensate for any imbalance in modulators 32 and 34.

A further consideration for providing the elimination of the even-order harmonic components concerns relative rf phase shift. Typically, the rf signals 36 and 38 are derived from the same rf source (not shown) split by an rf splitter (not shown). In order to effectively provide for the elimination of the even-order harmonic components created by the combining of the signals I_1 and I_2 and phase shifting the signals I_1 and I_2 with respect to each other, the accumulated rf phase shift encountered by a

signal traveling on a first path from the rf phase splitter to the rf combiner 52 by way of modulator 32 is preferably the same as the accumulated rf phase shift encountered by a signal traveling on a second path from the rf phase splitter to the rf combiner 52 by way of modulator 34. The first and second paths are of an of an electrical conductive medium from the rf splitter to the modulators 32 and 34, of an optical medium from the modulators 32 and 34 to the detectors 40 and 44, and of an electrical conductive medium from the detectors 40 and 44 to the combiner 52. The adaptation so that the accumulated rf phase shift is substantially the same in each of the first and second paths may be provided, in a manner known in the art, by adjusting the piezoelectric fiber stretcher 48.

It should now be appreciated that the practice of the present invention provides for an optical link that modulates an optical carrier signal with an information signal and does so without generating any unwanted even-order harmonic components so that the optical link has a multi-octave response characteristic.

In the practice of this invention, testing was performed and some of the results are illustrated in Figs. 5 and 6. The results shown in Fig. 5 were obtained using the prior art optical link 10 of Fig. 1 for $\alpha=\pi/4$, and the results shown in Fig. 6 were

obtained using the balanced detection optical link 28 of Fig. 3 of the present invention, also for $\alpha=\pi/4$. The results were obtained using an information signal having a frequency of 110 MHz.

5 Fig. 5 is composed of Figs. 5A and 5B each of which have an X axis given in frequency (2 kiloHertz/division (kHz/div)), and a Y axis illustrating an amplitude given in decibels (10dB/div). Fig. 5A illustrates a plot 56 of the amplitude of the fundamental component of the 110 MHz information signal, and Fig. 5B
10 illustrates a plot 58 of the amplitude of the second harmonic component. A comparison between Figs. 5A and 5B reveals that the degree of distortion related to the second harmonic component of the optical link 10 is relatively high.

 Fig. 6 is composed of Figs. 6A and 6B each having X and Y
15 axes respectively similar to Figs. 5A and 5B and with the same scales as Figs. 5A and 5B. Fig. 6A illustrates a plot 60 of the amplitude of the fundamental component of the 110 MHz information signal, whereas Fig. 6B indicates the plot 62 of the amplitude of the second harmonic component. A comparison between Figs. 6A and
20 6B reveals that the optical link 28 of Fig. 3 generates a low second harmonic component signal relative to the fundamental component. A further comparison between Figs. 5B and 6B reveals

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that the optical link 28 reduces the unwanted second harmonic component by a factor of about 28dB, representing a substantial improvement for the optical link 28 relative to the prior art optical link 10.

5 Further testing was performed in the practice of this invention using the prior art optical link 10 of Fig. 1 having the results shown in Fig. 7, and the optical link 28 of Fig. 3 having results shown in Fig. 8. Both Figs. 7 and 8 have an X axis indicating the applied rf signal given in dBm, and a Y axis
10 indicating the output signal P_{AC} of the respective links, and also given in dBm. Both Figs. 7 and 8 illustrate plots, indicated given by a reference number, for the performance parameters given in Table 1.

TABLE 1

REFERENCE NO.	PARAMETER
64	Input Thermal Noise
66	Noise Factor (NF)
68	Spurious Free Dynamic Range (SFDR)
70	Measured Output Noise
72	Plot of the Fundamental (110 MHz) Signal
74	Plot of the Second Order (220 MHz) Signal
76	Plot of Third Order (330 MHz) Signal

The associated current in which the related detector circuits 16 (Fig. 1) and 40 and 44 (Fig. 3) operated was approximately 1.1 milliamps thereby advantageously keeping each in its linear range of operation. The associated laser was a Nd:YAG source with an operating wavelength of 1.3 micrometers and supplied an optical carrier I_0 having a power of 300 milliwatts. The optical carrier I_0 was modulated by an information signal of 110 MHz.

A comparison between Figs. 7 and 8 reveals that a 28dB reduction in the second harmonic at -25dbm rf input (plots 74)

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was achieved using the optical link 28 of Fig. 3. Furthermore, a multi-octave spurious free dynamic range (sometimes referred to as linear dynamic range) (plots 68) was improved by 22dB, and the noise factor (plots 66) was improved by 3dB. Furthermore, the sensitivity of the single optical link 10 of Fig. 1 was measured and determined to be 0.15 microvolts, whereas the sensitivity of the optical link 28 of Fig. 3 was measured and determined to be 0.09 microvolts. This increased sensitivity makes the optical link 28 particularly suited for many electrical (E) field sensing application.

It should now be appreciated that the practice of the present invention provides for an optical link having an improved linear dynamic range, an improved sensitivity and a reduced noise factor, while at the same time having reduced unwanted even-order harmonic components.

Although the present invention has been primarily described as being used in telecommunication systems, the improved sensitivity of the optical link 28 also makes it particularly well suited to serve as an electric field sensor.

Furthermore, although the hereinbefore given description utilized a single laser source 12, the practice of the present

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invention contemplates that each of the modulators 32 and 34 of
Fig. 3 may be arranged to accept separate laser signals generated
by separate laser sources. This arrangement may be provided
because the optical phase related to the laser source does not
5 affect the quality of the combined output signal resulting in the
final output signal P_{Ac} of the optical link of the present
invention.

It should, therefore, be readily understood that many
modifications and variations of the present invention are
10 possible. It is,
therefore, to be understood that
the invention may be practiced otherwise than as
specifically described.

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ABSTRACT

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An optical link is disclosed that is operated by a balanced detection technique so as to extend its linear dynamic range, yet substantially eliminate even-order harmonic distortion of the modulated signal. The modulator comprises two parallel arranged Mach-Zehnder interferometers that respectively have a steady state DC voltage bias applied so that the operating points of the Mach-Zehnder interferometers are selected to produce two separate modulated signals, one of which is later phase shifted by a phase shift device so as to provide an output that is combined with the other modulated signal in such a manner so as to cause a subtraction, leaving the final output signal substantially free of any unwanted even-order harmonic components in order to provide for a multi-octave response characteristic.

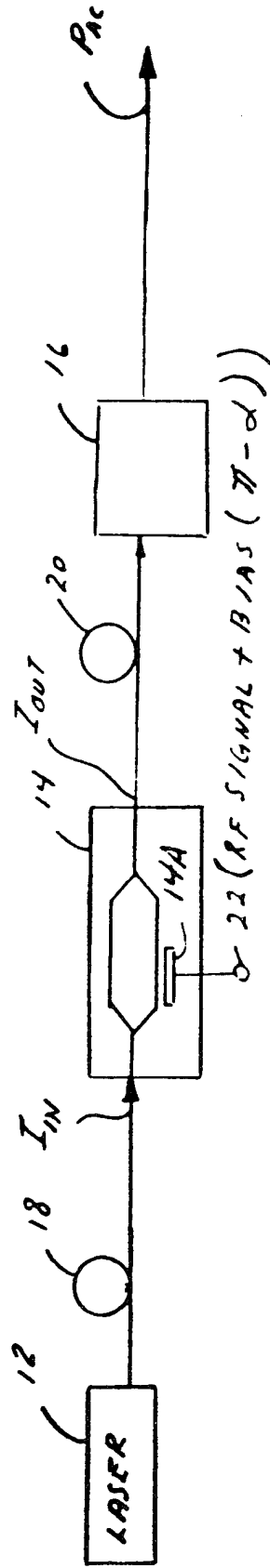


FIG. 1

FIG. 1 OR FIG. 2

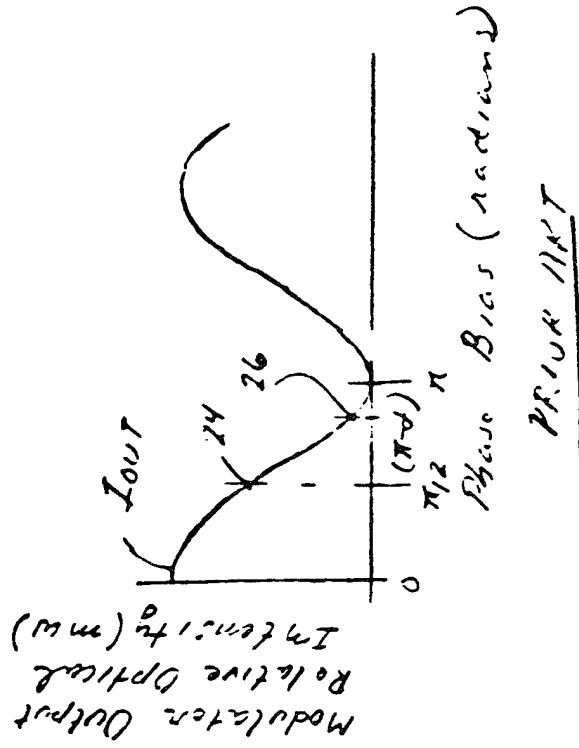


FIG. 2

FIG. 1 OR FIG. 2

10
8

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28

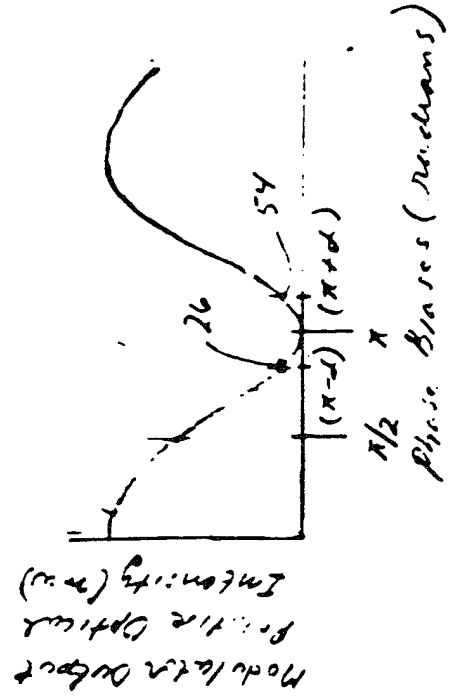
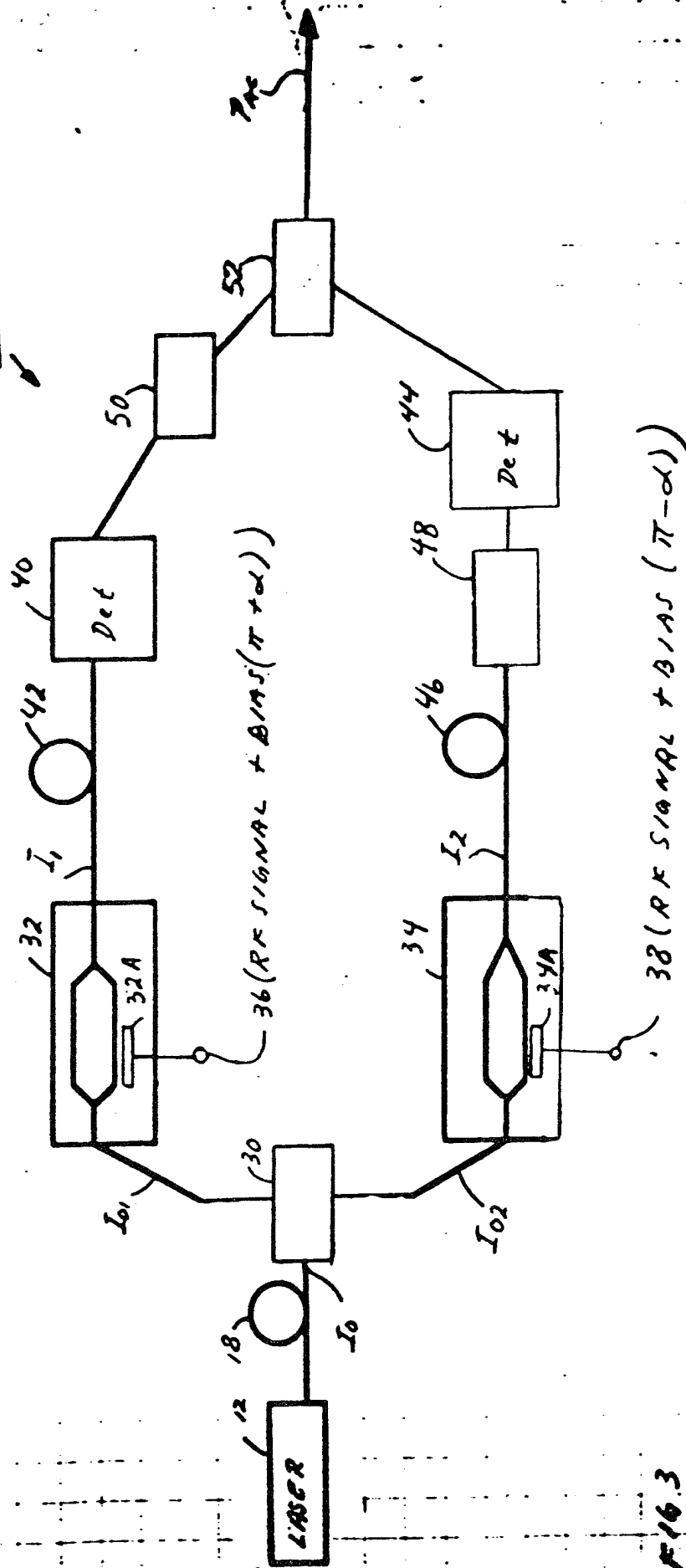


FIG 3

FIG 4

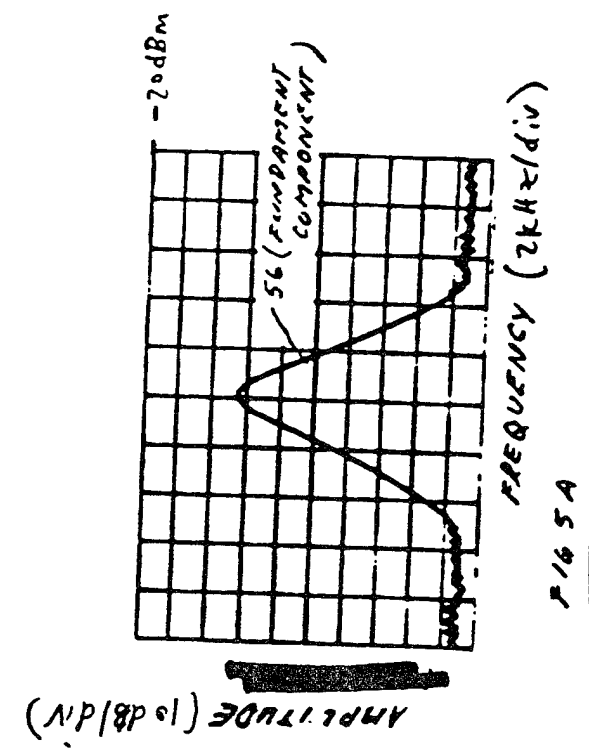


FIG 5 (PRIOR ART)

FIG 5A

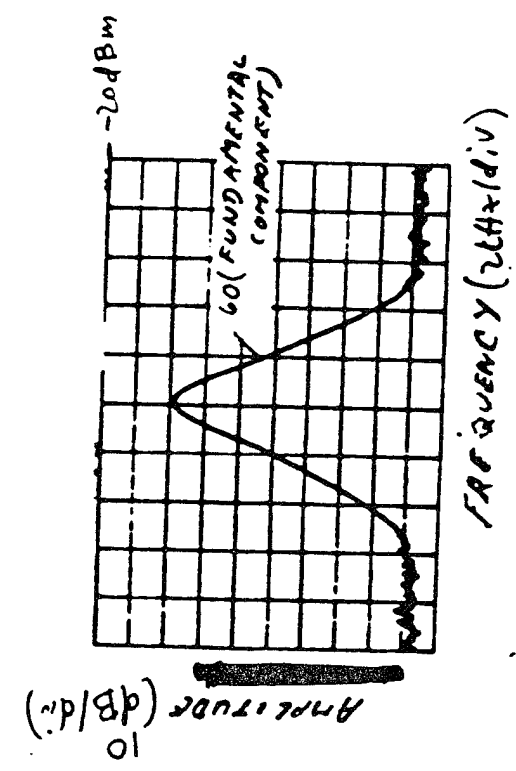


FIG 6A

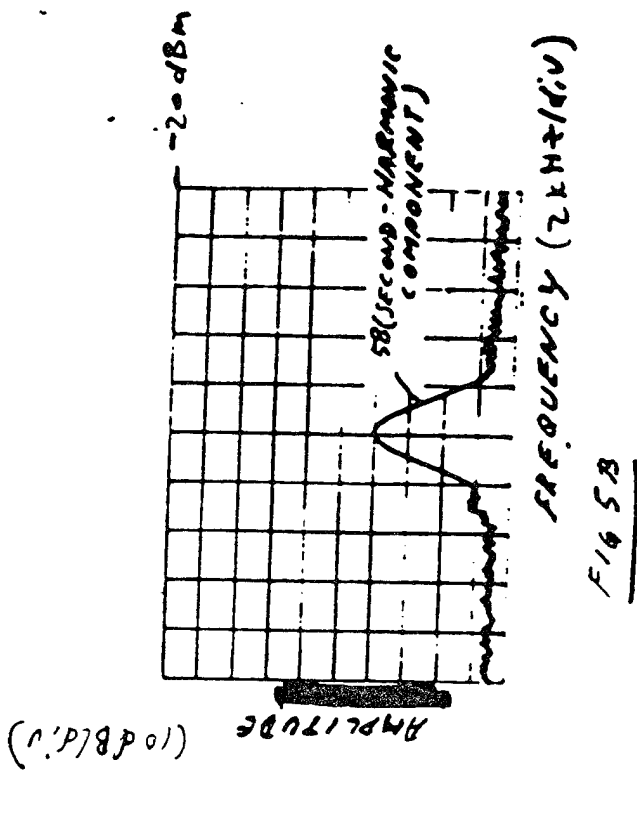


FIG 5B

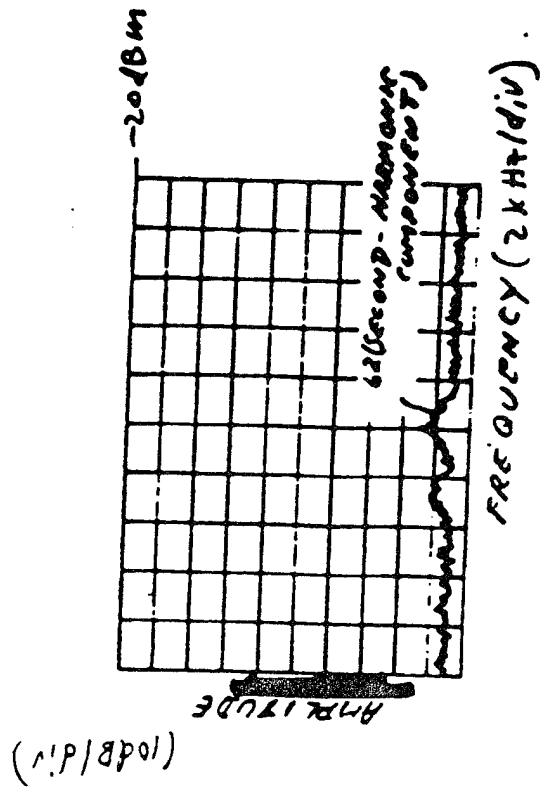


FIG 6B

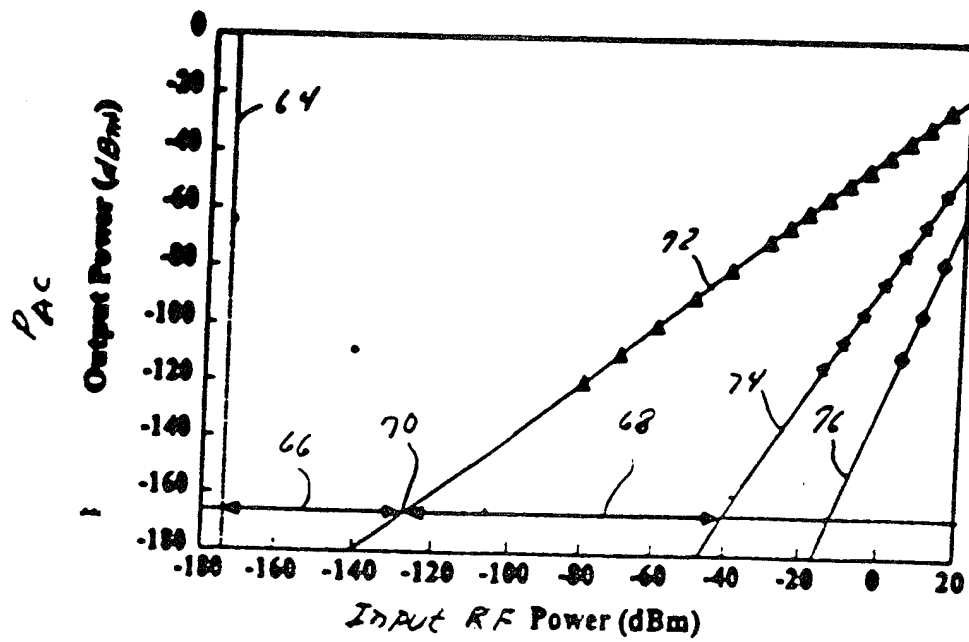


FIG. 7

PRIOR ART

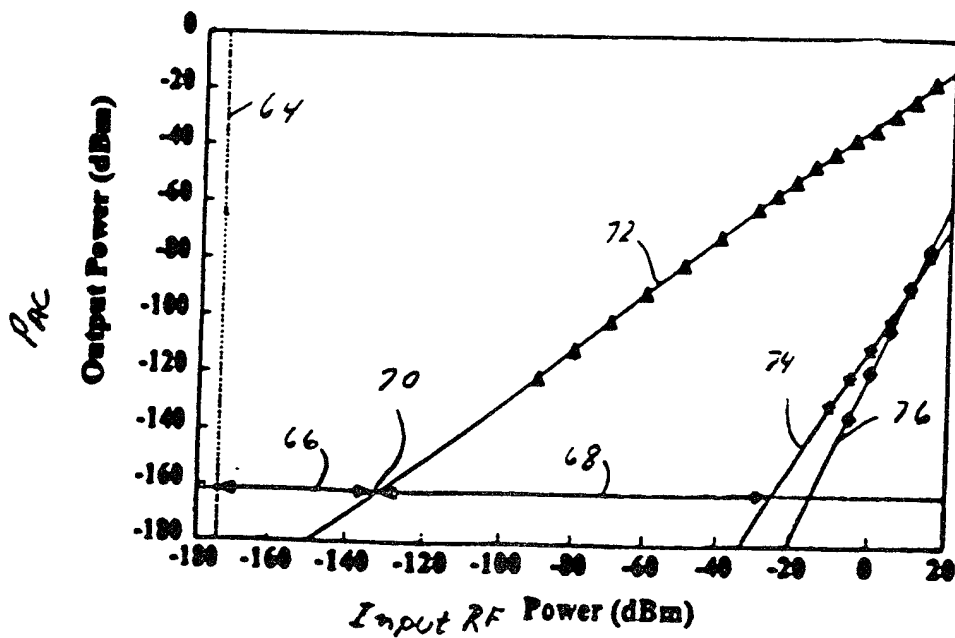


FIG. 8